

**PA96** 

RoHS

OMPLIA

# **Power Operational Amplifier**

### **FEATURES**

- High Voltage 300V
- High Output Current 1.5A
- 70 Watt Dissipation Capability
- 175 MHz Gain Bandwidth
- 250 V/µs Slew Rate

## **APPLICATIONS**

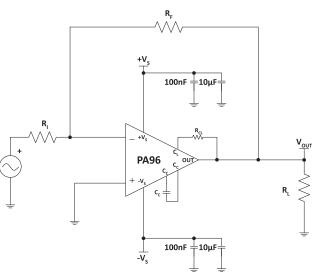
- PZT Drive
- Magnetic Deflection
- Programmable Power Supplies
- 70V Line Audio to 70W

## DESCRIPTION

The PA96 is a state of the art high voltage, high current operational amplifier designed to drive resistive, capacitive and inductive loads. For optimum linearity, the output stage is biased for class A/B operation. External compensation provides user flexibility in maximizing bandwidth at any gain setting. The safe operating area (SOA) can be observed for all operating conditions by selection of user programmable current limit. For continuous operation under load, a heatsink of proper rating is required.

The hybrid integrated circuit utilizes thick film (cermet) resistors, ceramic capacitors and semiconductor chips to maximize reliability, minimize size and give top performance. Ultrasonically bonded aluminum wires provide reliable interconnections at all operating temperatures. The 8-pin TO-3 package is hermetically sealed and electrically isolated. The use of compressible isolation washers voids the warranty.

## **TYPICAL CONNECTION**



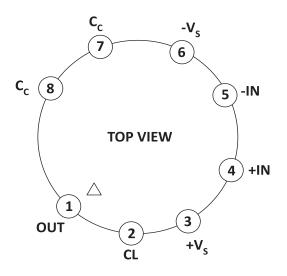
#### **Figure 1: Typical Connection**





## PINOUT AND DESCRIPTION TABLE

## Figure 2: External Connections



Pin Number	Name	Description	
1	OUT	The output. Connect this pin to load and to the feedback resistors.	
2	CL	Connect to the current limit resistor, and then the OUT pin. Output current fle into/out of this pin through R <sub>CL</sub> .	
3	+Vs	The positive supply rail.	
4	+IN	The non-inverting input.	
5	-IN	The inverting input.	
6	-Vs	The negative supply rail.	
7, 8	СС	Compensation capacitor connection. Select value based on Phase Compensation. See applicable section.	



## **SPECIFICATIONS**

The power supply voltage specified under typical (TYP) applies unless noted as a test condition.

### ABSOLUTE MAXIMUM RATINGS

Parameter	Symbol	Min	Max	Units
Supply Voltage, total	+V <sub>s</sub> to -V <sub>s</sub>		300	V
Output Current, source, continuous	Ι <sub>Ο</sub>		1.5	A
Power Dissipation, internal, DC	P <sub>D</sub>		70	W
Input Voltage, differential	V <sub>IN (Diff)</sub>		±15	V
Input Voltage, common mode	V <sub>cm</sub>	-V <sub>S</sub>	+V <sub>S</sub>	V
Temperature, pin solder, 10s max.			350	°C
Temperature, junction <sup>1</sup>	Tj		150	°C
Temperature Range, storage		-65	150	°C
Operating Temperature Range, case	T <sub>C</sub>	-55	125	°C

1. Long term operation at the maximum junction temperature will result in reduced product life. Derate power dissipation to achieve high MTTF.

#### CAUTION

The internal substrate contains beryllia (BeO). Do not break the seal. If accidentally broken, do not crush, machine, or subject to temperatures in excess of 850°C to avoid generating toxic fumes.

#### INPUT

Parameter	Test Conditions	Min	Тур	Max	Units
Offset Voltage			1	5	mV
Offset Voltage vs. temperature	Full temp range		20	50	μV/°C
Offset Voltage vs. supply				20	μ٧/٧
Bias Current, initial				200	pА
Bias Current vs. Supply				0.1	pA/V
Offset Current, initial				50	рА
Input Resistance, DC			10 <sup>11</sup>		Ω
Input Capacitance			4		pF
Common Mode Voltage Range <sup>1</sup>		±V <sub>S</sub> -13			V
Common Mode Rejection, DC		92			dB
Noise	100 kHz bandwidth, 1 k $\Omega$ $R_S$		6		μV RMS

1.  $+V_S$  and  $-V_S$  denote the positive and negative supply rail respectively. Total  $V_S$  is measured from  $+V_S$  to  $-V_S$ .



# PA96

## GAIN

Parameter	Test Conditions	Min	Тур	Max	Units
Open Loop @ 15 Hz	R <sub>L</sub> = 1 kΩ, C <sub>C</sub> =100pF	96	114		dB
Gain Bandwidth Product @ 1 MHz	Vs = 150V, -Vs = 150V, A = - 100, RF = 100 kΩ	100	175		MHz
Phase Margin	Full temp range, using recom- mended C <sub>C</sub> for gain.	60			o
PBW	250V p-p output, 100 Ω, +150V Supplies, Cc = 0pf		100		kHz

## OUTPUT

Parameter	Test Conditions	Min	Тур	Max	Units
Voltage Swing <sup>1</sup>	I <sub>O</sub> = 1.5A	+V <sub>S</sub> - 12	+V <sub>S</sub> - 5.6		V
Voltage Swing <sup>1</sup>	I <sub>O</sub> = -1.5A	-V <sub>S</sub> + 12	-V <sub>S</sub> + 10		V
Voltage Swing <sup>1</sup>	I <sub>O</sub> = 0.1A	+V <sub>S</sub> - 8			V
Voltage Swing <sup>1</sup>	I <sub>O</sub> = -0.1A	-V <sub>S</sub> + 8			V
Current, continuous, DC		1.5			А
Slew Rate	$A_V = -100$ , ±150V Supplies, 250 Ω load negative slope, Positive slope much faster	200	250		V/µs
Settling Time, to 0.1%	A <sub>V</sub> = -100, 1V Step, C <sub>C</sub> = 0pF		2		μs
Resistance, open loop	DC, 1A Load		7	10	Ω

1. +V<sub>S</sub> and -V<sub>S</sub> denote the positive and negative supply rail respectively. Total V<sub>S</sub> is measured from +V<sub>S</sub> to -V<sub>S</sub>.

#### **POWER SUPPLY**

Parameter	Test Conditions	Min	Тур	Max	Units
Voltage		±15	±100	±150	V
Current, Quiescent total		25	30	35	mA
Current, Quiescent output stage only			10		mA



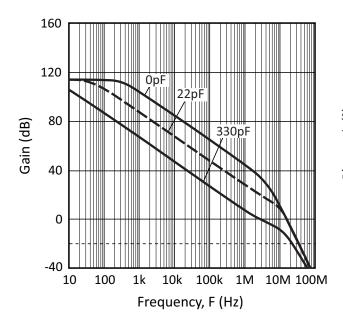
#### THERMAL

Parameter	Test Conditions	Min	Тур	Max	Units
Resistance, AC Junction to Case <sup>1</sup>	Full temp range, f > 60 Hz		1.2	1.3	°C/W
Resistance, DC Junction to Case	Full temp range, f < 60 Hz		1.6	1.8	°C/W
Resistance, Junction to Ambient			30		°C/W
Temperature Range, case	Meets full range specs	-25		85	°C

1. Rating applies if the output current alternates between both output transistors at a rate faster than 60 Hz.



## **TYPICAL PERFORMANCE GRAPHS**



## Figure 3: Open Loop Frequency Response

Figure 5: Gain Bandwidth vs. +Supply Voltage

## Figure 4: Open Loop Phase Response

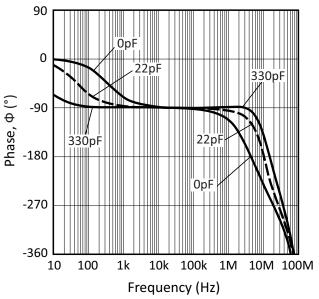
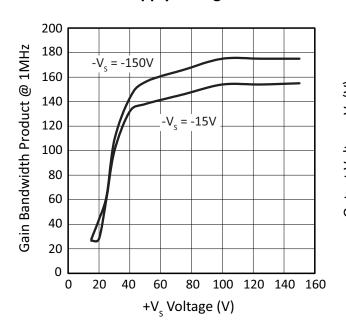


Figure 6: Rail to Rail Pulse Response



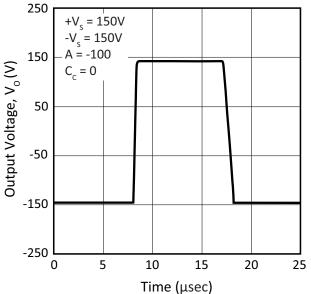




Figure 8: Large Signal Pulse Response

125

75

+V<sub>s</sub> = 150V -V<sub>s</sub> = 150V

A = -100

 $C_c = 0$ 

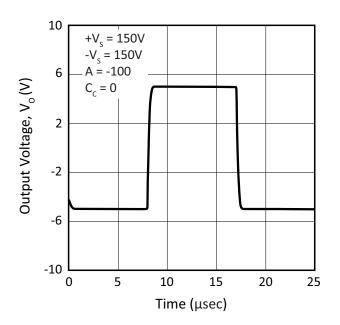


Figure 9: Output Voltage Swing

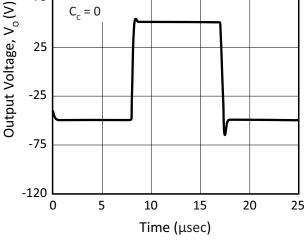
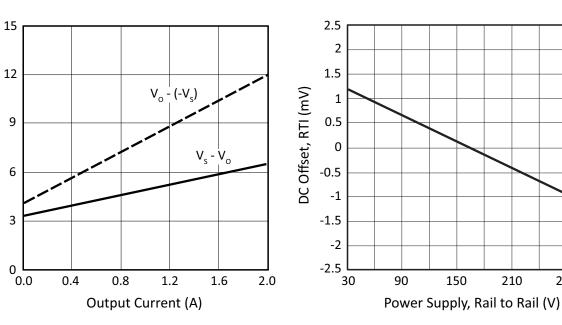


Figure 10: DC Offset vs. Power Supply

210

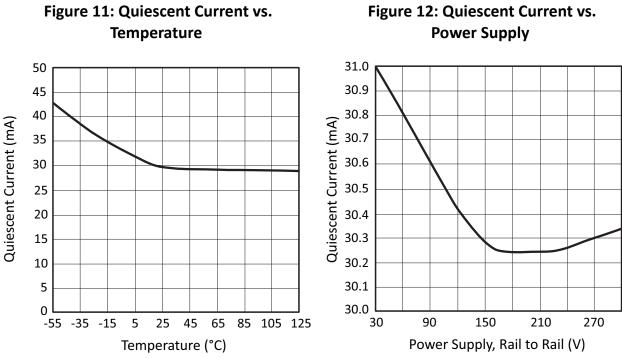
270



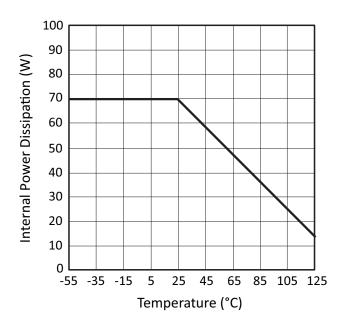


|V<sub>s</sub>-V<sub>o</sub>| (V)





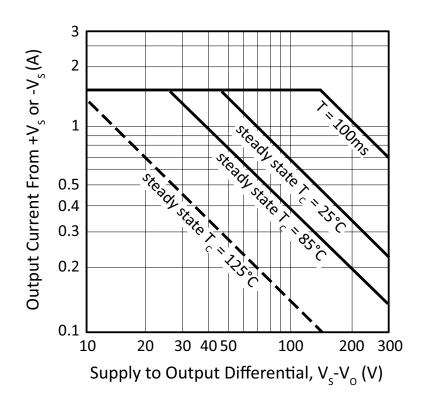






SAFE OPERATING AREA (SOA)







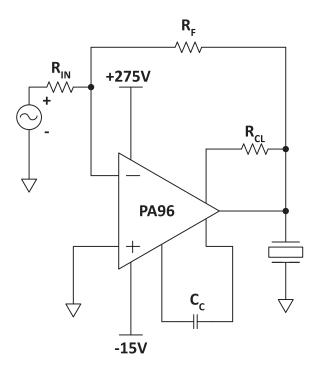
## GENERAL

Please read Application Note 1 "General Operating Considerations" which covers stability, supplies, heat sinking, mounting, current limit, SOA interpretation, and specification interpretation. Visit www.apexanalog.com for Apex Microtechnology's complete Application Notes library, Technical Seminar Workbook, and Evaluation Kits.

## **TYPICAL APPLICATION**

The MOSFET output stage of the PA96 provides superior SOA performance compared to bipolar output stages where secondary breakdown is a concern. The extended SOA is ideal in applications where the load is highly reactive and may impose simultaneously both high voltage and high current across the output stage transistors. In the figure above a piezo-electric transducer is driven to high currents and high voltages by the PA96.

Figure 15: Typical Application (PZT Position Control)



## **SPECIAL PRECAUTIONS**

The PA96 operates with up 300V rail to rail voltage, and delivers up to 1.5A of current. Precautions should be taken for the safety of the user and the amplifier.

Although the non-operating common mode input range is rail to rail, the differential input voltage must not exceed ±15 V.

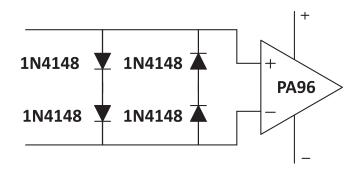
Therefore; if the feedback ratio is less than 10, even if caused by disconnecting a signal source, typical power turn on transients can destroy the amplifier.

Similarly in a voltage follower application a large differential transient can be generated if the slew rate of the input is greater than that of the voltage follower.



Therefore it is prudent to clamp the input with series back to back diodes as shown below.





If experimentally optimizing the compensation capacitor, turn off the supplies and let them bleed to low voltage before installing each new value. Otherwise internal current pulses of up to 3 amps can be induced. Also, do you want your fingers around 300V?

Essentially the full rail to rail power supply voltage may be applied to the compensation capacitor. A 400V COG or Mica capacitor is recommended.

## POWER BANDWIDTH

The power bandwidth is  $1/(\pi x \text{ the negative edge slew time})$ . The slew time is determined by the compensation capacitor, load, and internal device capacitance; it is independent of closed loop gain. The uncompensated power bandwidth is typically 100 kHz for a 250Vp-p output signal into 100  $\Omega$ . It typically increases to above 300 kHz with no load.

## **COMPENSATION TABLE**

The following table tabulates recommended compensation capacitor values vs. gain. These values will typically result in less than 2% overshoot and a -3db small signal bandwidth of greater than 1 MHz, except under operating conditions where uncompensated gain bandwidth is too low to support a 1 MHz bandwidth. (See Gain Bandwidth vs. Plus Power Supply curves). Note that other factors such as capacitance in parallel with the feedback resistor may reduce circuit bandwidth from that determined from the gain bandwidth curve.

Cc	Inverting Gain			
	From	То		
150pf	1	2		
51pf	2	5		
33pf	5	10		
22pf	10	20		
10pf	20	50		
5pf	50	100		
None	100	up		



Cc	Non-Inverting Gain			
	From	То		
330pf	1	2		
150pf	2	3		
51pf	3	6		
33pf	6	10		
22pf	10	20		
10pf	20	50		
5pf	50	100		
None	100	up		

## **CURRENT LIMIT**

For proper operation the current limit resistor,  $R_{CL}$ , must be connected as shown in the external connections diagram. The minimum value is 0.2  $\Omega$ , with a maximum practical value of 100  $\Omega$ . For optimum reliability the resistor should be set as high as possible. The value is calculated as:

$$I_L(A) = \frac{0.68V}{R_{CL}(\Omega)}$$

Note that the 0.68V is reduced by 2mV every °C rise in temperature.

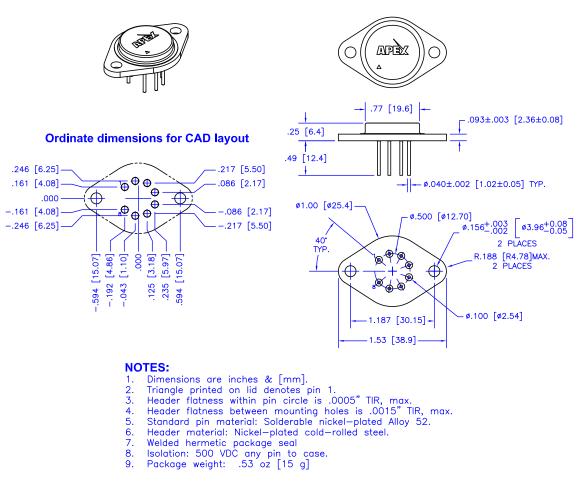
Also note that the current limit can be set such that the SOA is exceeded on a continuous basis. As an example if the current limit was set at 1.5A and the supply was at 150V, a short to ground would produce 225 watts internal dissipation, greatly exceeding the 83 watt steady state SOA rating.

Under some conditions of load and compensation the amplifier may oscillate at a low level when current limit is active, even though the amplifier is stable otherwise. The current will be limited to the programmed value in this situation. To minimize such occurrences, use a non-reactive resistor to program current limit.



#### PACKAGE OPTIONS

#### PACKAGE STYLE CE



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